MOTOROLA

MC14568B

CMOS MSI

(LOW-POWER COMPLEMENTARY MOS)

PHASE COMPARATOR AND PROGRAMMABLE COUNTERS

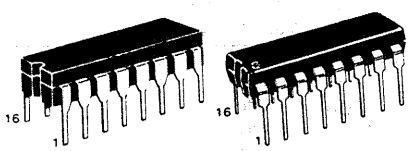
PHASE COMPARATOR AND PROGRAMMABLE COUNTERS

The MC14568B consists of a phase comparator, a divide-by-4, 16, 64 or 100 counter and a programmable divide-by-N 4-bit binary counter (all positive-edge triggered) constructed with MOS P-channel and N-channel enhancement mode devices (complementary MOS) in a single monolithic structure.

The MC14568B has been designed for use in conjunction with a programmable divide-by-N counter for frequency synthesizers and phase-locked loop applications requiring low power dissipation and/or high noise immunity.

This device can be used with both counters cascaded and the output of the second counter connected to the phase comparator (CTL high), or used separate of the programmable divide-by-N counter, for example cascaded with MC14569B (CTL low), MC14522B or MC14526B.

- Quiescent Current = 5.0 nA typ/pkg @ 5 Vdc
- Supply Voltage Range = 3.0 Vdc to 18 Vdc
- Capable of Driving Two Low-Power TTL Loads, One Low-Power Schottky TTL Load or Two HTL Loads Over the Rated Temperature Range.



L SUFFIX
CERAMIC PACKAGE
CASE 620

P SUFFIX
PLASTIC PACKAGE
CASE 648

ORDERING INFORMATION

MC14XXXB — Suffix Denotes
- L Ceramic Package
- P Plastic Package
- A Extended Operating Temperature Range
- C Limited Operating Temperature Range

MAXIMUM RATINGS (Voltages referenced to V_{SS})

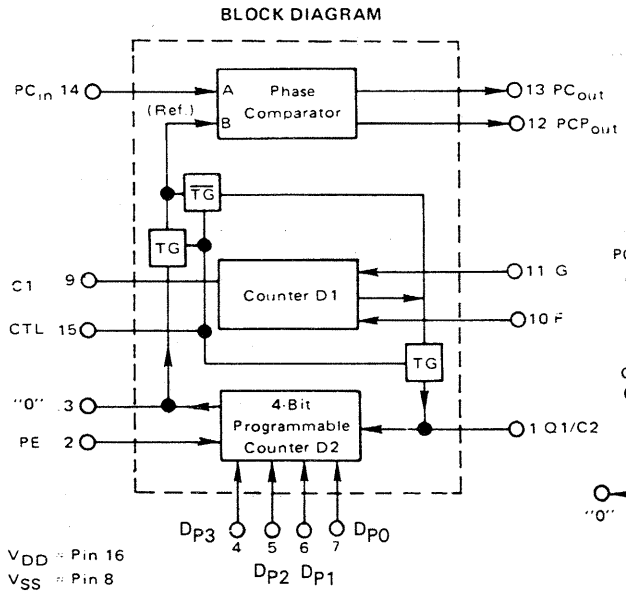
Rating	Symbol	Value	Unit
DC Supply Voltage	V_{DD}	-0.5 to +18	Vdc
Input Voltage, All Inputs	V_{in}	-0.5 to V_{DD} + 0.5	Vdc
DC Current Drain per Pin	I	10	mAdc
Operating Temperature Range — AL Device CL/CP Device	T_A	-55 to +125 -40 to +85	°C
Storage Temperature Range	T_{stg}	-65 to +150	°C

TRUTH TABLE

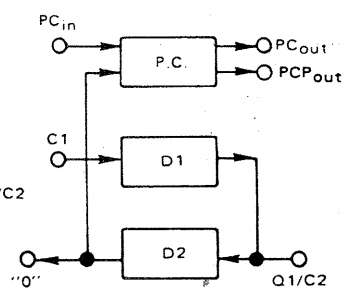
F Pin 10	G Pin 11	Division Ratio of Counter D1
0	0	4
0	1	16
1	0	64
1	1	100

The divide-by-zero state on the programmable divide-by-N 4-bit binary counter, D2, is illegal.

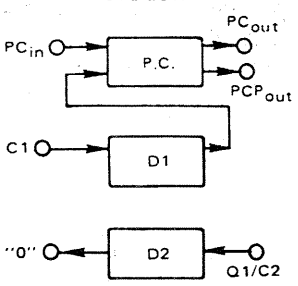
BLOCK DIAGRAM



CTL HIGH



CTL LOW



ELECTRICAL CHARACTERISTICS

Characteristic	Symbol	V_{DD} Vdc	T_{low}* Min	T_{low}* Max	25°C Min	25°C Typ	25°C Max	T_{high}* Min	T_{high}* Max	Unit
Output Voltage "0" Level	V_{OL}	5.0	–	0.05	–	0	0.05	–	0.05	Vdc
V_{in} = V_{DD} or 0		10	–	0.05	–	0	0.05	–	0.05	
		15	–	0.05	–	0	0.05	–	0.05	
"1" Level	V_{OH}	5.0	4.95	–	4.95	5.0	–	4.95	–	Vdc
V_{in} = 0 or V_{DD}		10	9.95	–	9.95	10	–	9.95	–	
		15	14.95	–	14.95	15	–	14.95	–	
Input Voltage#† "0" Level	V_{IL}									Vdc
(V_O = 4.5 or 0.5 Vdc)		5.0	–	1.5	–	2.25	1.5	–	1.5	
(V_O = 9.0 or 1.0 Vdc)		10	–	3.0	–	4.50	3.0	–	3.0	
(V_O = 13.5 or 1.5 Vdc)		15	–	4.0	–	6.75	4.0	–	4.0	
"1" Level	V_{IH}									
(V_O = 0.5 or 4.5 Vdc)		5.0	3.5	–	3.5	2.75	–	3.5	–	Vdc
(V_O = 1.0 or 9.0 Vdc)		10	7.0	–	7.0	5.50	–	7.0	–	
(V_O = 1.5 or 13.5 Vdc)		15	11.0	–	11.0	8.25	–	11.0	–	
Output Drive Current (AL Device)	I_{OH}									mAdc
(V_{OH} = 2.5 Vdc) Source		5.0	–1.2	–	–1.0	–1.7	–	–0.7	–	
(V_{OH} = 4.6 Vdc)		5.0	–0.25	–	–0.2	–0.36	–	–0.14	–	
(V_{OH} = 9.5 Vdc)		10	–0.62	–	–0.5	–0.9	–	–0.35	–	
(V_{OH} = 13.5 Vdc)		15	–1.8	–	–1.5	–3.5	–	–1.1	–	
(V_{OL} = 0.4 Vdc) Sink	I_{OL}	5.0	0.64	–	0.51	0.88	–	0.36	–	mAdc
(V_{OL} = 0.5 Vdc)		10	1.6	–	1.3	2.25	–	0.9	–	
(V_{OL} = 1.5 Vdc)		15	4.2	–	3.4	8.8	–	2.4	–	
Output Drive Current (CL/CP Device)	I_{OH}									mAdc
(V_{OH} = 2.5 Vdc) Source		5.0	–1.0	–	–0.8	–1.7	–	–0.6	–	
(V_{OH} = 4.6 Vdc)		5.0	–0.2	–	–0.16	–0.36	–	–0.12	–	
(V_{OH} = 9.5 Vdc)		10	–0.5	–	–0.4	–0.9	–	–0.3	–	
(V_{OH} = 13.5 Vdc)		15	–1.4	–	–1.2	–3.5	–	–1.0	–	
(V_{OL} = 0.4 Vdc) Sink	I_{OL}	5.0	0.52	–	0.44	0.88	–	0.36	–	mAdc
(V_{OL} = 0.5 Vdc)		10	1.3	–	1.1	2.25	–	0.9	–	
(V_{OL} = 1.5 Vdc)		15	3.6	–	3.0	8.8	–	2.4	–	
Input Current (AL Device)	I_{in}	15	–	±0.1	–	±0.00001	±0.1	–	±1.0	µAdc
Input Current (CL/CP Device)	I_{in}	15	–	±0.3	–	±0.00001	±0.3	–	±1.0	µAdc
Input Capacitance (V_{in} = 0)	C_{in}	–	–	–	–	5.0	7.5	–	–	pF
Quiescent Current (AL Device) (Per Package)	I_{DD}	5.0	–	5.0	–	0.005	5.0	–	150	µAdc
		10	–	10	–	0.010	10	–	300	
		15	–	20	–	0.015	20	–	600	
Quiescent Current (CL/CP Device) (Per Package)	I_{DD}	5.0	–	20	–	0.005	20	–	150	µAdc
		10	–	40	–	0.010	40	–	300	
		15	–	80	–	0.015	80	–	600	
Total Supply Current**† (Dynamic plus Quiescent, Per Package) (C_L = 50 pF on all outputs, all buffers switching)	I_T	5.0	I_T = (0.2 µA/kHz) f + I_{DD}							µAdc
		10	I_T = (0.4 µA/kHz) f + I_{DD}							
		15	I_T = (0.9 µA/kHz) f + I_{DD}							
Three-State Leakage Current, Pins 1, 13 (AL Device)	I_{TL}	15	–	±0.1	–	±0.00001	±0.1	–	±3.0	µAdc
Three-State Leakage Current, Pins 1, 13 (CL/CP Devices)	I_{TL}	15	–	±1.0	–	±0.00001	±1.0	–	±7.5	µAdc

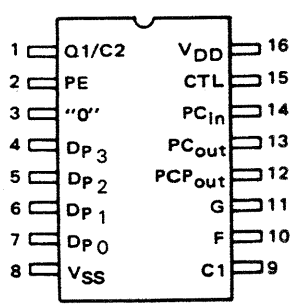
*T_{low} = -55°C for AL Device, -40°C for CL/CP Device.
T_{high} = +125°C for AL Device, +85°C for CL/CP Device.
#Noise immunity specified for worst-case input combination.
Noise Margin for both "1" and "0" level = 1.0 Vdc min @ V_{DD} = 5.0 Vdc
2.0 Vdc min @ V_{DD} = 10 Vdc
2.5 Vdc min @ V_{DD} = 15 Vdc

†To calculate total supply current at loads other than 50 pF:
$I_T(C_L) = I_T(50\ pF) + 1 \times 10^{-3} (C_L - 50) V_{DD} f$
where: I_T is in µA (per package), C_L in pF, V_{DD} in Vdc, and f in kHz is input frequency.

**The formulas given are for the typical characteristics only at 25°C.

†Pin 15 is connected to V_{SS} or V_{DD} for input voltage test.

PIN ASSIGNMENT



SWITCHING CHARACTERISTICS (C_L = 50 pF, T_A = 25°C)

Characteristic	Symbol	V_{DD} Vdc	Min	Typ	Max	Unit
Output Rise Time	t_{TLH}	5.0	–	180	360	ns
		10	–	90	180	
		15	–	65	130	
Output Fall Time	t_{THL}	5.0	–	100	200	ns
		10	–	50	100	
		15	–	40	80	
Minimum Pulse Width, C1, Q1/C2, or PC_{in} Input	t_{WH}	5.0	–	125	250	ns
		10	–	60	120	
		15	–	45	90	
Maximum Clock Rise and Fall Time, C1, Q1/C2, or PC_{in} Input	t_{TLH}, t_{THL}	5.0	15	–	–	μs
		10	15	–	–	
		15	15	–	–	
PHASE COMPARATOR						
Input Resistance	Rin	5.0 to 15	–	10^6	–	MΩ
Input Sensitivity, DC Coupled	–	5.0 to 15	See Input Voltage			
Turn-Off Delay Time, PC_{out} and PCP_{out} Outputs	t_{PHL}	5.0	–	550	1100	ns
		10	–	195	390	
		15	–	120	240	
Turn-On Delay Time, PC_{out} and PCP_{out} Outputs	t_{PHL}	5.0	–	675	1350	ns
		10	–	300	600	
		15	–	190	380	
DIVIDE-BY-4, 16, 64 OR 100 COUNTER (D1)						
Maximum Clock Pulse Frequency	f_{cl}					MHz
Division Ratio = 4, 64 or 100		5.0	3.0	6.0	–	
		10	8.0	16	–	
		15	10	22	–	
Division Ratio = 16		5.0	1.0	2.5	–	MHz
		10	3.0	6.3	–	
		15	5.0	9.7	–	
Propagation Delay Time, Q1/C2 Output	t_{PLH}, t_{PHL}					ns
Division Ratio = 4, 64 or 100		5.0	–	450	900	
		10	–	190	380	
		15	–	130	260	
Division Ratio = 16		5.0	–	720	1440	ns
		10	–	300	600	
		15	–	200	400	
PROGRAMMABLE DIVIDE-BY-N 4-BIT COUNTER (D2)						
Maximum Clock Pulse Frequency (Figure 3a)	f_{cl}	5.0	1.2	1.8	–	MHz
		10	3.0	8.5	–	
		15	4.0	12	–	
Turn-On Delay Time, "0" Output (Figure 3a)	t_{PLH}	5.0	–	450	900	ns
		10	–	190	380	
		15	–	130	260	
Turn-Off Delay Time, "0" Output (Figure 3a)	t_{PHL}	5.0	–	225	450	ns
		10	–	85	170	
		15	–	60	150	
Minimum Preset Enable Pulse Width	$t_{WH(PE)}$	5.0	–	75	250	ns
		10	–	40	100	
		15	–	30	75	

This device contains circuitry to protect the inputs against damage due to high static voltages or electric fields: however, it is advised that normal precautions be taken to avoid application of any voltage higher than maximum rated voltages to this high impedance circuit. For proper operation it is recommended that V_{in} and V_{out} be constrained to the range $V_{SS} \leq (V_{in}$ or $V_{out}) \leq V_{DD}$.
Unused inputs must always be tied to an appropriate logic voltage level (e.g. either V_{SS} or V_{DD}).

SWITCHING TIME TEST CIRCUITS AND WAVEFORMS

FIGURE 1 – PHASE COMPARATOR

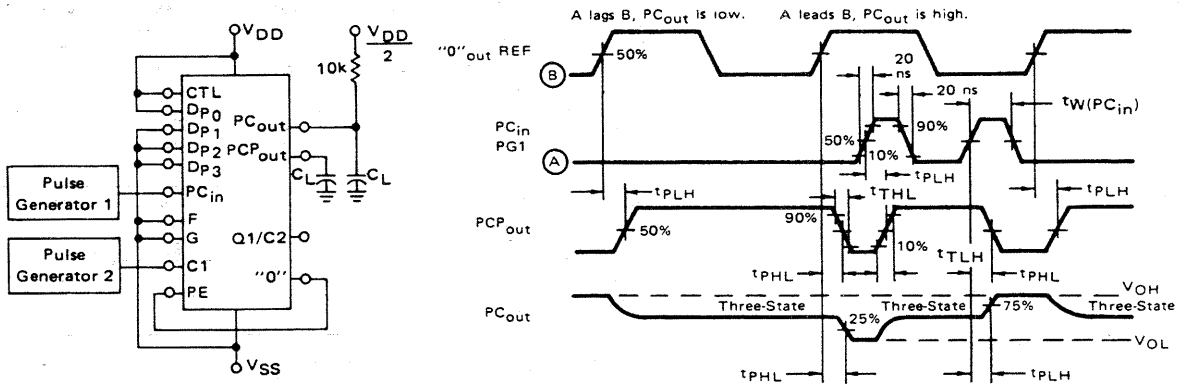


FIGURE 2 – COUNTER D1

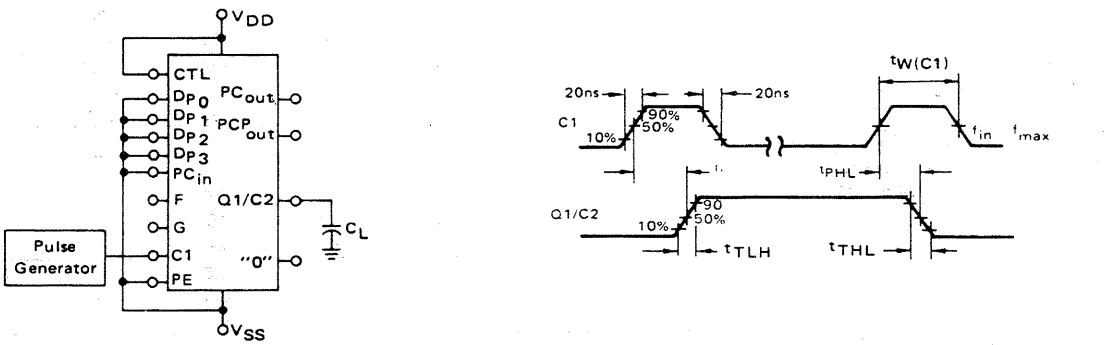
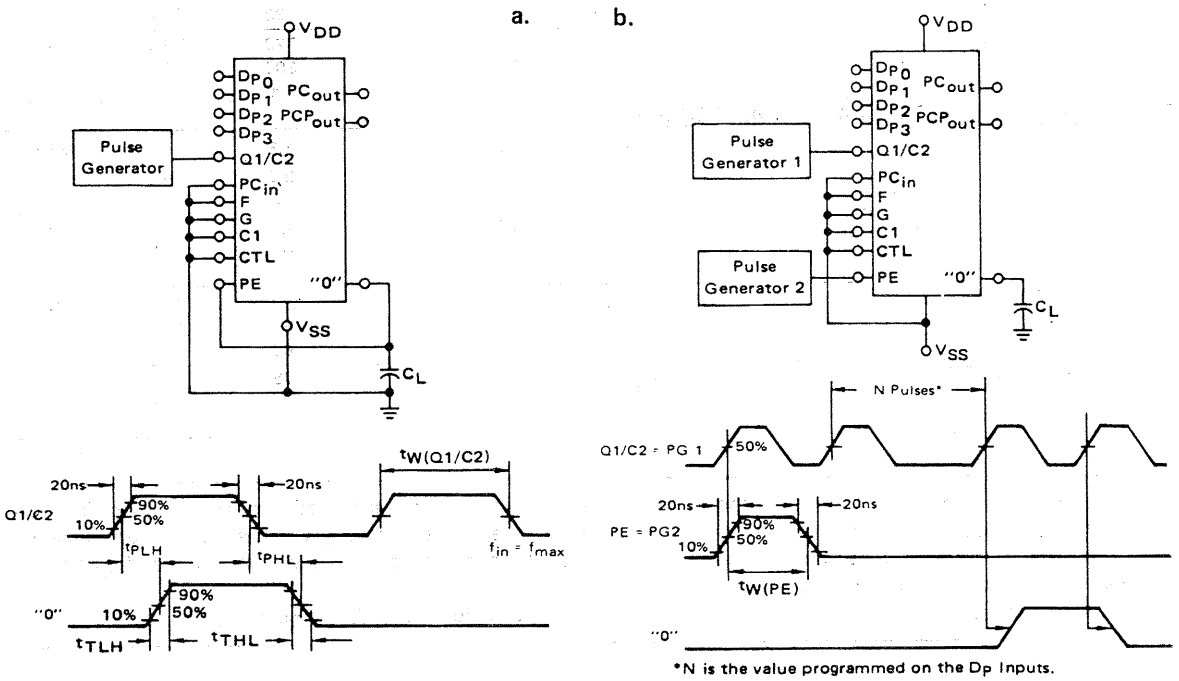
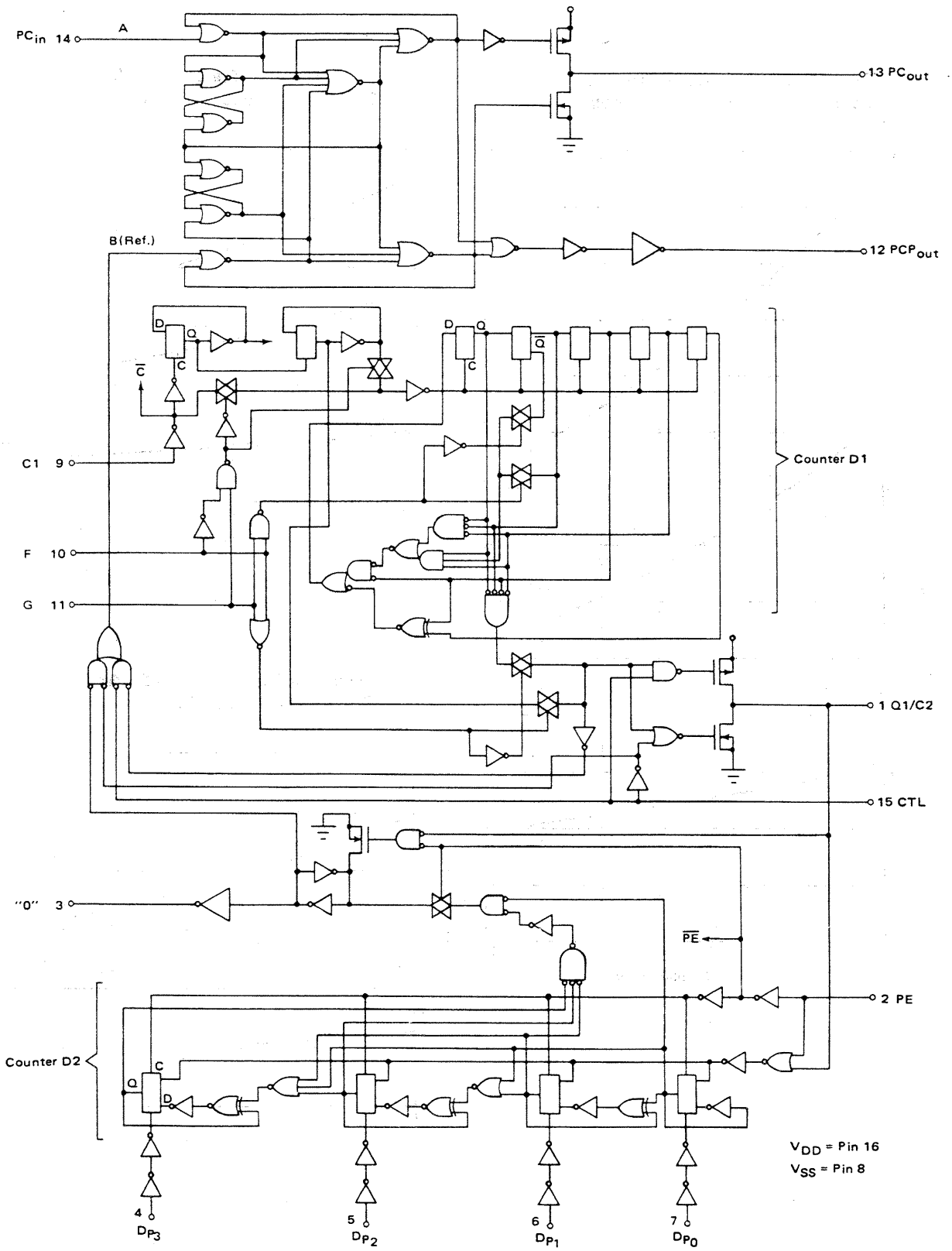


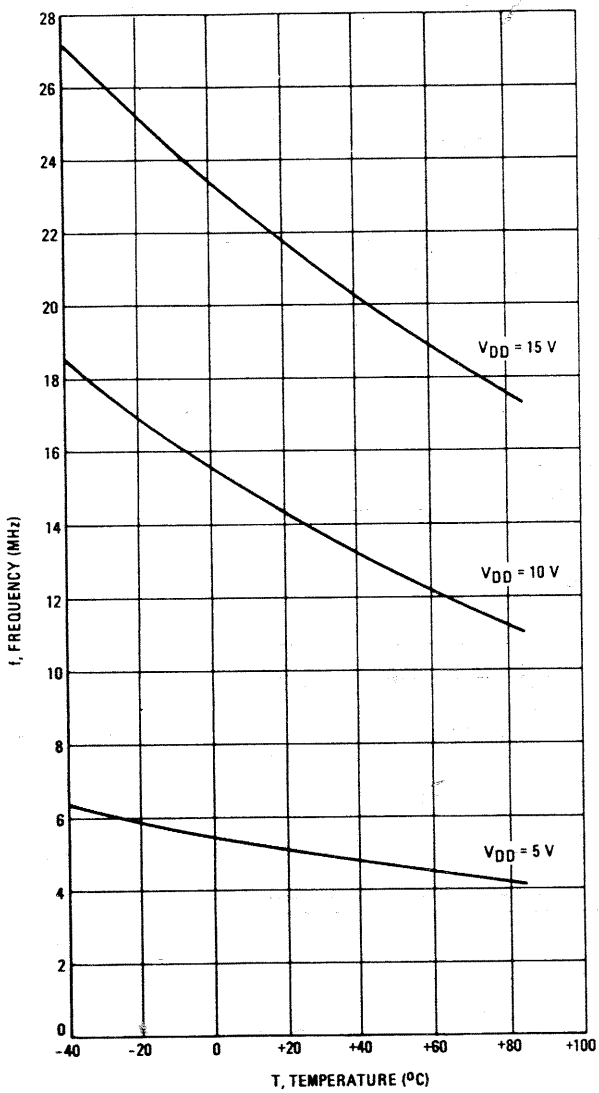
FIGURE 3 – COUNTER D2



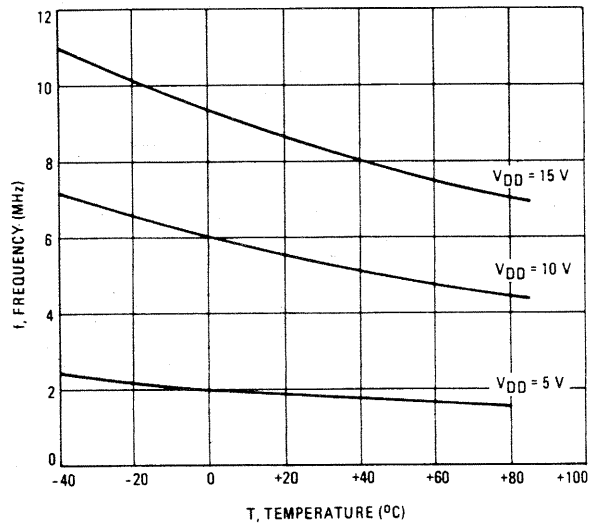
LOGIC DIAGRAM



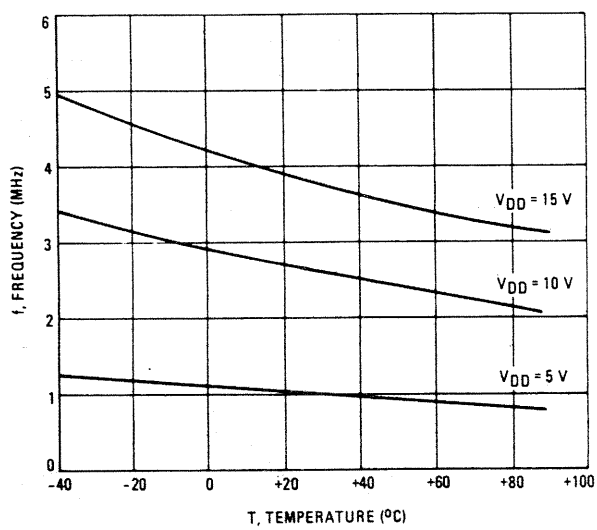
Typical Maximum Frequency Divider D1
Division ratios: 4, 64 or 100 (CL = 50 pF)



Typical Maximum Frequency Divider D1
Division ratio: 16 (CL = 50 pF)



Typical Maximum Frequency Divider D2
Division ratio: 2 (CL = 50 pF)



OPERATING CHARACTERISTICS

The MC14568B contains a phase comparator, a fixed divider (÷ 4, ÷ 16, ÷ 64, ÷ 100) and a programmable divide-by-N 4-bit counter.

PHASE COMPARATOR

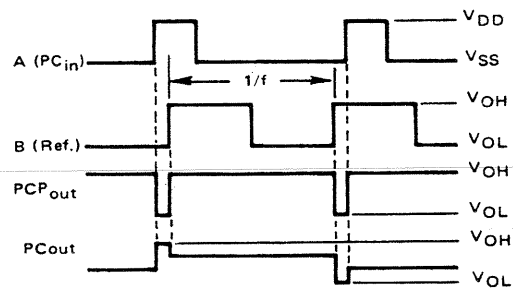
The phase comparator is a positive edge controlled logic circuit. It essentially consists of four flip-flops and an output pair of MOS transistors. Only one of its inputs (PC_{in}, pin 14) is accessible externally. The second is connected to the output of one of the two counters D1 or D2 (see block diagram).

Duty cycles of both input signals (at A and B) need not be taken into consideration since the comparator responds to leading edges only.

If both input signals have identical frequencies but different phases, with signal A (pin 14) leading signal B (Ref.), the comparator output will be high for the time equal to the phase difference.

If signal A lags signal B, the output will be low for the same time. In between, the output will be in a three-state condition and the voltage on the capacitor of an RC filter normally connected at this point will have some intermediate value (see Figure 4). When used in a phase locked loop, this value will adjust the Voltage Controlled Oscillator frequency by reducing the phase difference between the reference signal and the divided VCO frequency to zero.

FIGURE 4 – PHASE COMPARATOR WAVEFORMS



If the input signals have different frequencies, the output signal will be high when signal B has a lower frequency than signal A, and low otherwise.

Under the same conditions of frequency difference, the output will vary between V_{OH} (or V_{OL}) and some intermediate value until the frequencies of both signals are equal and their phase difference equal to zero, i.e. until locked condition is obtained.

Capture and lock range will be determined by the VCO frequency range. The comparator is provided with a lock indicator output, which will stay at logic 1 in locked conditions.

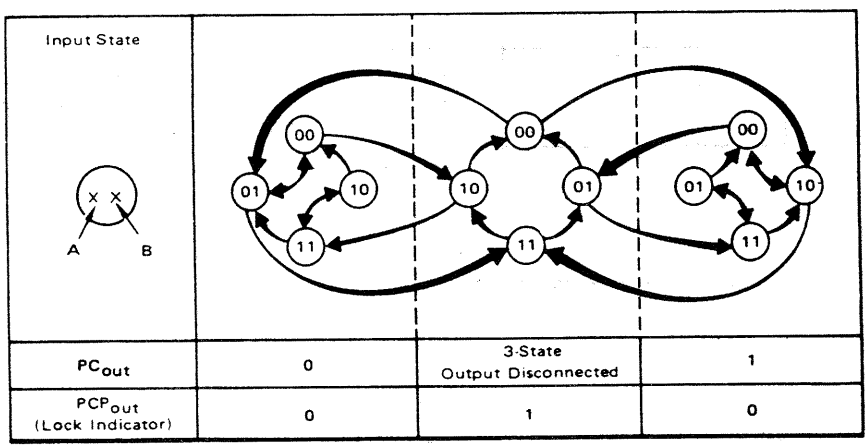
The state diagram (Figure 5) depicts the internal state transitions. It assumes that only one transition on either signal occurs at any time. It shows that a change of the output state is always associated with a positive transition of either signal. For a negative transition, the output does not change state. A positive transition may not cause the output to change; this happens when the signals have different frequencies.

DIVIDE BY 4, 16, 64 OR 100 COUNTER (D1)

This counter is able to work at an input frequency of 5 MHz for a V_{DD} value of 10 volts over the standard temperature range when dividing by 4, 64 and 100. Programming is accomplished by use of inputs F and G (pins 10 and 11) according to the truth table shown. Connecting the Control input (CTL, pin 15) to V_{DD} allows cascading this counter with the programmable divide-by-N counter provided in the same package. Independent operation is obtained when the Control input is connected to V_{SS}.

The different division ratios have been chosen to generate the reference frequences corresponding to the channel spacings normally required in frequency synthesizer applications. For example, with the division ratio 100 and a 5 MHz crystal stabilized source a reference frequency of 50 kHz is supplied to the comparator. The lower division ratios permit operation with low frequency crystals.

FIGURE 5 – PHASE COMPARATOR STATE DIAGRAM



PC_{out}	0	3-State Output Disconnected	1
PCP_{out} (Lock Indicator)	0	1	0

If used in cascade with the programmable divide-by-N counter, practically all usual reference frequencies, or channel spacings of 25, 20, 12.5, 10, 6.25 kHz, etc. are easily achievable.

PROGRAMMABLE DIVIDE-BY-N 4-BIT COUNTER (D2)

This counter is programmable by using inputs D_{P0} . . . D_{P3} (pins 7 . . . 4). The Preset Enable input enables the parallel preset inputs D_{P0} . . . D_{P3}. The "0" output must be externally connected to the PE input for single stage applications. Since there is not a cascade feedback input, this counter, when cascaded, must be used as the most significant digit. Because of this, it can be cascaded with binary counters as well as with BCD counters (MC14569B, MC14522B, MC14526B).

TYPICAL APPLICATIONS

FIGURE 6 – CASCADING MC14568B AND MC14522B OR MC14526B WITH MC14569B

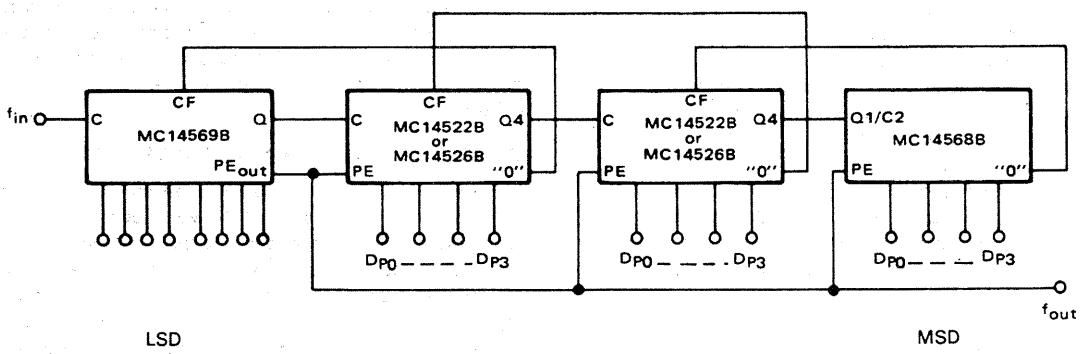
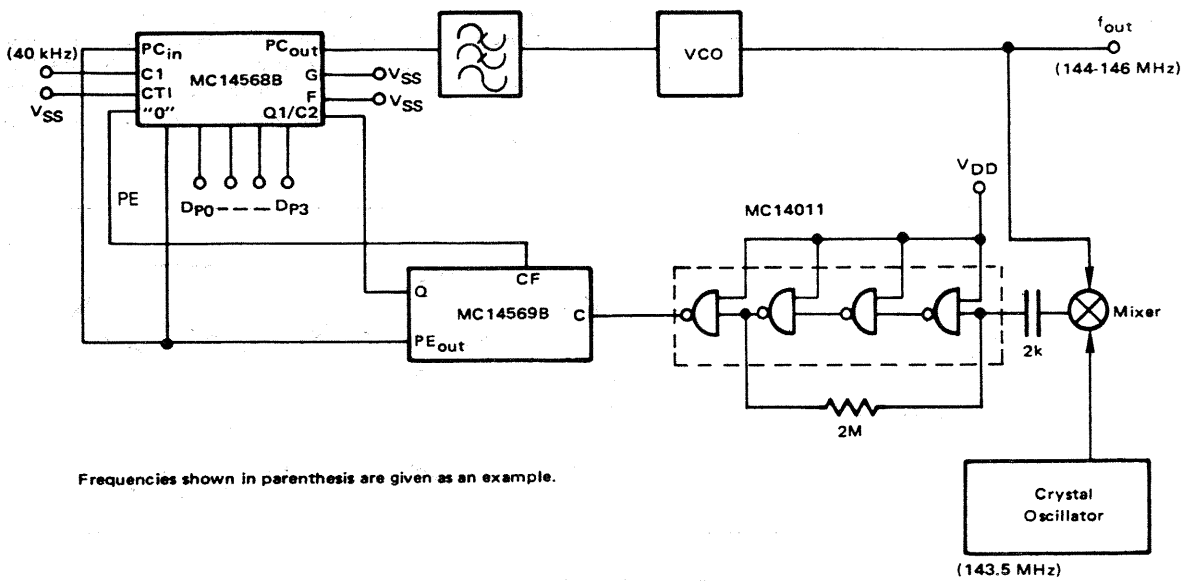
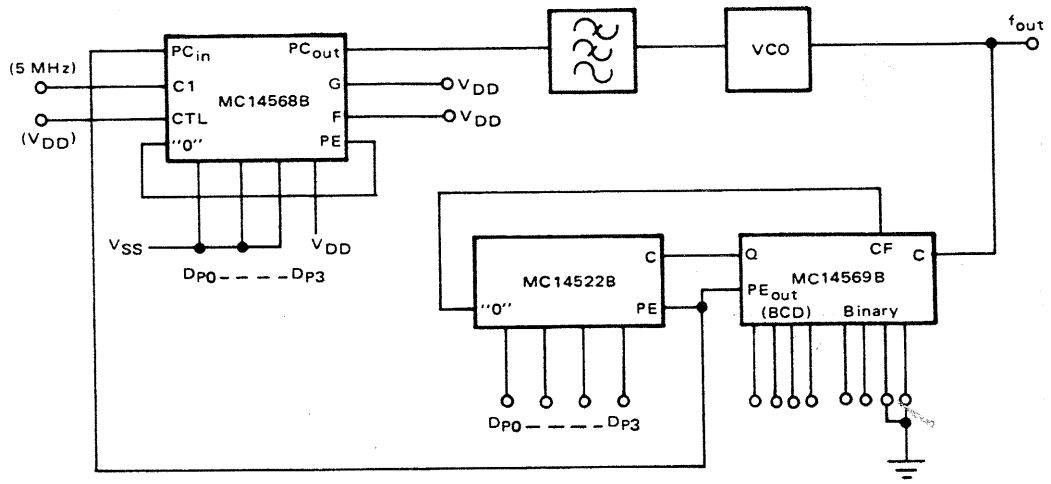


FIGURE 7 – FREQUENCY SYNTHESIZER WITH MC14568B and MC14569B USING A MIXER
(Channel Spacing 10 kHz)



Frequencies shown in parenthesis are given as an example.

FIGURE 8 – FREQUENCY SYNTHESIZER USING MC14568B, MC14569B AND MC14522B
(Without Mixer)



N_1
(0–5)
(625 kHz steps)

N_2
(0–9)
(62.5 kHz steps)

N_3
(0, 4, 8, 12)
(6.25 kHz steps)

Divide ratio = $160N_1 + 16N_2 + N_3$

Example:
$f_{out} = N_1$ (MHz) + N_2 (x100 kHz) + N_3 (x25 kHz)
Frequency range = 5 MHz
Channel spacing = 25 kHz
Reference frequency = 6.25 kHz

Figures shown in parenthesis refer to example.

FIGURE 9 – TYPICAL 23-CHANNEL CB FREQUENCY SYNTHESIZER FOR DOUBLE CONVERSION TRANSCEIVERS